

Small, Low Power, 3-Axis ±3 g iMEMS® Accelerometer

ADXL330

FEATURES

3-axis sensing Small, low-profile package 4 mm × 4 mm × 1.45 mm LFCSP Low power 200 μA at V₅ = 2.0 V (typical) Single-supply operation 2.0 V to 3.6 V 10,000 g shock survival Excellent temperature stability BW adjustment with a single capacitor per axis RoHS/WEEE lead-free compliant

APPLICATIONS

Cost-sensitive, low power, motion- and tilt-sensing applications Mobile devices Gaming systems

- **Disk drive protection**
- Image stabilization
- Sports and health devices

GENERAL DESCRIPTION

The ADXL330 is a small, thin, low power, complete three axis accelerometer with signal conditioned voltage outputs, all on a single monolithic IC. The product measures acceleration with a minimum full-scale range of ± 3 g. It can measure the static acceleration of gravity in tilt-sensing applications, as well as dynamic acceleration resulting from motion, shock, or vibration.

The user selects the bandwidth of the accelerometer using the C_X , C_Y , and C_Z capacitors at the X_{OUT} , Y_{OUT} , and Z_{OUT} pins. Bandwidths can be selected to suit the application, with a range of 0.5 Hz to 1,600 Hz for X and Y axes, and a range of 0.5 Hz to 550 Hz for the Z axis.

The ADXL330 is available in a small, low-profile, 4 mm \times 4 mm \times 1.45 mm, 16-lead, plastic lead frame chip scale package (LFCSP_LQ).



FUNCTIONAL BLOCK DIAGRAM

Rev. 0

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SPECIFICATIONS

 $T_A = 25^{\circ}C$, $V_S = 3 V$, $C_X = C_Y = C_Z = 0.1 \mu F$, acceleration = 0 g, unless otherwise noted. All minimum and maximum specifications are guaranteed. Typical specifications are not guaranteed.

Table 1.					
Parameter	Conditions	Min	Тур	Мах	Unit
SENSOR INPUT	Each axis				
Measurement Range		±3	±3.6		g
Nonlinearity	% of full scale		±0.3		%
Package Alignment Error			±1		Degrees
Inter-Axis Alignment Error			±0.1		Degrees
Cross Axis Sensitivity ¹			±1		%
SENSITIVITY (RATIOMETRIC) ²	Each axis				
Sensitivity at Xout, Yout, Zout	$V_S = 3 V$	270	300	330	mV/g
Sensitivity Change Due to Temperature ³	$V_S = 3 V$		±0.015		%/°C
ZERO g BIAS LEVEL (RATIOMETRIC)	Each axis				
0 g Voltage at Хоит, Yоит, Zouт	$V_S = 3 V$	1.2	1.5	1.8	V
0 g Offset vs. Temperature			±1		mg∕°C
NOISE PERFORMANCE					
Noise Density Xout, Yout			280		µg/√Hz rms
Noise Density Zout			350		µg/√Hz rms
FREQUENCY RESPONSE ⁴					
Bandwidth Xout, Yout⁵	No external filter		1600		Hz
Bandwidth Zout⁵	No external filter		550		Hz
R _{FILT} Tolerance			32 ± 15%		kΩ
Sensor Resonant Frequency			5.5		kHz
SELF-TEST ⁶					
Logic Input Low			+0.6		V
Logic Input High			+2.4		V
ST Actuation Current			+60		μΑ
Output Change at Xout	Self-test 0 to 1		-150		mV
Output Change at Yout	Self-test 0 to 1		+150		mV
Output Change at Zout	Self-test 0 to 1		-60		mV
OUTPUT AMPLIFIER					
Output Swing Low	No load		0.1		V
Output Swing High	No load		2.8		V
POWER SUPPLY					
Operating Voltage Range		2.0		3.6	V
Supply Current	$V_S = 3 V$		320		μΑ
Turn-On Time ⁷	No external filter		1		ms
TEMPERATURE					
Operating Temperature Range		-25		+70	°C

¹ Defined as coupling between any two axes.

² Sensitivity is essentially ratiometric to V_s.

³ Defined as the output change from ambient-to-maximum temperature or ambient-to-minimum temperature.

⁴ Actual frequency response controlled by user-supplied external filter capacitors (C_x, C_Y, C_Z).

⁵ Bandwidth with external capacitors = $1/(2 \times \pi \times 32 \text{ k}\Omega \times \text{C})$. For C_x, C_Y = 0.003 µF, bandwidth = 1.6 kHz. For C_z = 0.01 µF, bandwidth = 500 Hz. For C_x, C_Y, C_z = 10 µF, bandwidth = 1.6 kHz. For C_z = 0.01 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 1.6 kHz. For C_z = 0.01 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 1.6 kHz. For C_z = 0.01 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 1.6 kHz. For C_z = 0.01 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.01 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 1.6 kHz. For C_z = 0.01 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x, C_Y = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 Hz. For C_x = 0.003 µF, bandwidth = 500 bandwidth = 0.5 Hz.

⁶ Self-test response changes cubically with Vs. ⁷ Turn-on time is dependent on Cx, Cy, Cz and is approximately $160 \times C_x$ or Cy or Cz + 1 ms, where Cx, Cy, Cz are in μ F.

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APPLICATIONS POWER SUPPLY DECOUPLING

For most applications, a single 0.1 μ F capacitor, C_{DC}, placed close to the ADXL330 supply pins adequately decouples the accelerometer from noise on the power supply. However, in applications where noise is present at the 50 kHz internal clock frequency (or any harmonic thereof), additional care in power supply bypassing is required as this noise can cause errors in acceleration measurement. If additional decoupling is needed, a 100 Ω (or smaller) resistor or ferrite bead can be inserted in the supply line. Additionally, a larger bulk bypass capacitor (1 μ F or greater) can be added in parallel to C_{DC}. Ensure that the connection from the ADXL330 ground to the power supply ground is low impedance because noise transmitted through ground has a similar effect as noise transmitted through V_s.

SETTING THE BANDWIDTH USING C_x, C_y, AND C_z

The ADXL330 has provisions for band limiting the X_{OUT} , Y_{OUT} , and Z_{OUT} pins. Capacitors must be added at these pins to implement low-pass filtering for antialiasing and noise reduction. The equation for the 3 dB bandwidth is

$$F_{-3 \text{ dB}} = 1/(2\pi(32 \text{ k}\Omega) \times C_{(X, Y, Z)})$$

or more simply

 $F_{-3 \text{ dB}} = 5 \ \mu F / C_{(X, Y, Z)}$

The tolerance of the internal resistor (R_{FILT}) typically varies as much as $\pm 15\%$ of its nominal value (32 k Ω), and the bandwidth varies accordingly. A minimum capacitance of 0.0047 μ F for C_x, C_y, and C_z is recommended in all cases.

Bandwidth (Hz)	Capacitor (µF)
1	4.7
10	0.47
50	0.10
100	0.05
200	0.027
500	0.01

Table 5. Filter Capacitor Selection, Cx, Cy, and Cz

SELF-TEST

The ST pin controls the self-test feature. When this pin is set to V_s, an electrostatic force is exerted on the accelerometer beam. The resulting movement of the beam allows the user to test if the accelerometer is functional. The typical change in output is -500 mg (corresponding to -150 mV) in the X-axis, 500 mg (or 150 mV) on the Y-axis, and -200 mg (or -60 mV) on the Z-axis. This ST pin may be left open circuit or connected to common (COM) in normal use.

Never expose the ST pin to voltages greater than V_s + 0.3 V. If this cannot be guaranteed due to the system design (for

instance, if there are multiple supply voltages), then a low V_F clamping diode between ST and V_S is recommended.

DESIGN TRADE-OFFS FOR SELECTING FILTER CHARACTERISTICS: THE NOISE/BW TRADE-OFF

The selected accelerometer bandwidth ultimately determines the measurement resolution (smallest detectable acceleration). Filtering can be used to lower the noise floor to improve the resolution of the accelerometer. Resolution is dependent on the analog filter bandwidth at X_{OUT}, Y_{OUT}, and Z_{OUT}.

The output of the ADXL330 has a typical bandwidth of greater than 500 Hz. The user must filter the signal at this point to limit aliasing errors. The analog bandwidth must be no more than half the analog-to-digital sampling frequency to minimize aliasing. The analog bandwidth can be further decreased to reduce noise and improve resolution.

The ADXL330 noise has the characteristics of white Gaussian noise, which contributes equally at all frequencies and is described in terms of $\mu g/\sqrt{Hz}$ (the noise is proportional to the square root of the accelerometer bandwidth). The user should limit bandwidth to the lowest frequency needed by the application to maximize the resolution and dynamic range of the accelerometer.

With the single-pole, roll-off characteristic, the typical noise of the ADXL330 is determined by

rms Noise = Noise Density
$$\times (\sqrt{BW \times 1.6})$$

Often, the peak value of the noise is desired. Peak-to-peak noise can only be estimated by statistical methods. Table 6 is useful for estimating the probabilities of exceeding various peak values, given the rms value.

lable 6. Esti	mation of	Peak-to-F	Peak Noise
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Peak-to-Peak Value	% of Time that Noise Exceeds Nominal Peak-to-Peak Value
$2 \times rms$	32
$4 \times rms$	4.6
6 × rms	0.27
8 × rms	0.006

USE WITH OPERATING VOLTAGES OTHER THAN 3 V

The ADXL330 is tested and specified at $V_s = 3$ V; however, it can be powered with V_s as low as 2 V or as high as 3.6 V. Note that some performance parameters change as the supply voltage is varied.

The ADXL330 output is ratiometric, therefore, the output sensitivity (or scale factor) varies proportionally to the supply voltage. At V_s = 3.6 V, the output sensitivity is typically 360 mV/g. At V_s = 2 V, the output sensitivity is typically 195 mV/g.

The zero *g* bias output is also ratiometric, so the zero *g* output is nominally equal to $V_s/2$ at all supply voltages.

The output noise is not ratiometric but is absolute in volts; therefore, the noise density decreases as the supply voltage increases. This is because the scale factor (mV/g) increases while the noise voltage remains constant. At V_s = 3.6 V, the X- and Y-axis noise density is typically 230 µg/ \sqrt{Hz} , while at V_s = 2 V, the X- and Y-axis noise density is typically 350 µg/ \sqrt{Hz} .

Self-test response in g is roughly proportional to the square of the supply voltage. However, when ratiometricity of sensitivity is factored in with supply voltage, the self-test response in volts is roughly proportional to the cube of the supply voltage. For example, at V_s = 3.6 V, the self-test response for the ADXL330 is approximately –275 mV for the X-axis, +275 mV for the Y-axis, and –100 mV for the Z-axis.

At $V_S = 2$ V, the self-test response is approximately -60 mV for the X-axis, +60 mV for the Y-axis, and -25 mV for the Z-axis.

The supply current decreases as the supply voltage decreases. Typical current consumption at V_s = 3.6 V is 375 µA, and typical current consumption at V_s = 2 V is 200 µA.



Figure 31. Axes of Acceleration Sensitivity, Corresponding Output Voltage Increases When Accelerated Along the Sensitive Axis



Figure 32. Output Response vs. Orientation to Gravity